

Lecture 27: Concept of feedback in amplifiers: Negative and Positive

Feedback in amplifiers

DI02011011: Electronics Circuit and Application (ECA)

Lecture Agenda & Outline

- 1. Universal Feedback Architecture: Open-Loop Amplifier, Feedback Network, and Summing Junctions
- 2. Closed-Loop Gain Derivation: Mathematical Formulation of $A_f = A / (1 + A * \beta)$ and Loop Gain T
- 3. Negative Feedback Desensitization: Immunity to Parameter Variations, Temperature Drift, and Aging
- 4. Performance Modifications: Bandwidth Extension, Gain-Bandwidth Invariance, and Harmonic Distortion Reduction
- 5. Positive Feedback Mechanics: Regenerative Action, Latching, and Barkhausen Stability Criterion ($-A * \beta = 1$, phase = 0 deg)
- 6. Quantitative Engineering Design Examples and Comparative Performance Analysis Matrix

General Feedback System Model & Block Diagram Topology

- System Components: A feedback amplifier consists of a basic open-loop forward amplifier with transfer function $A(s)$, a feedback network with transfer function $\beta(s)$, a sampling network, and a mixing/summing junction.
- Signal Variables: Input signal x_s (voltage or current), feedback signal $x_f = \beta * x_o$, net input signal $x_i = x_s - x_f$ (for negative feedback) or $x_i = x_s + x_f$ (for positive feedback).
- Forward Path Amplifier (A): Idealized open-loop gain $A = x_o / x_i$. Internal parameters are sensitive to temperature, DC bias shifts, and manufacturing tolerances.
- Feedback Network (β): Passive resistive/reactive attenuation network $\beta = x_f / x_o$ (typically $\beta \approx 1$). Designed with precise, temperature-stable passive components (resistors/capacitors).
- Mixer/Summing Node: Combines source signal x_s and feedback signal x_f to generate error/effective drive signal x_i .

Derivation of Closed-Loop Gain A_f & Loop Gain Dynamics

- Algebraic Derivation for Negative Feedback: Output equation $x_o = A x_i$. *Substituting net input $x_i = x_s - x_f$ gives $x_o = A (x_s - \beta x_o)$.*
- Rearranging Terms: $x_o + A \beta x_o = A x_s \Rightarrow x_o (1 + A \beta) = A x_s$.
- Closed-Loop Gain Formula: $A_f = x_o / x_s = A / (1 + A \beta)$.
- Loop Gain (T): The product $T = A \beta$ represents the total gain around the feedback loop (from mixer through forward path and feedback network back to mixer).
- Amount of Feedback (F): Defined as $F = 1 + A \beta = 1 + T$. *Expressed in decibels as $F_{dB} = 20 \log_{10}(1 + A \beta)$.*
- Asymptotic Gain Limit: Under heavy negative feedback (where $A \beta \gg 1$): $A_f \approx A / (A \beta) = 1 / \beta$! Closed-loop gain becomes virtually independent of open-loop gain A.

Mechanisms of Negative Feedback: Desensitization of Gain

- Sensitivity Definition: Fractional change in closed-loop gain A_f relative to fractional change in open-loop gain A : $S_{A^f} = (dA_f / A_f) / (dA / A)$.
- Derivation of Sensitivity Reduction: Differentiating A_f with respect to A gives $dA_f / dA = d/dA [A / (1 + A \beta)] = 1 / (1 + A \beta)^2$.
- Sensitivity Equation: $dA_f / A_f = (1 / (1 + A \beta)) (dA / A)$. Fractional gain variation is reduced by the factor $(1 + A \beta)$.
- Thermal & Aging Immunity: Temperature-induced variations in BJT β_0 or MOSFET g_m (which cause 50% - 100% drift in open-loop gain A) cause negligible drift ($\approx 1\%$) in closed-loop gain A_f .
- Nonlinear Stabilization: Linearizes overall amplifier transfer characteristics, even when transistor open-loop gain A fluctuates across large signal swings.

Bandwidth Extension via Negative Feedback

- High-Frequency Response with Feedback: Single-pole open-loop amplifier $A(s) = A_0 / (1 + s / \omega_H)$, where A_0 is open-loop midband gain and $\omega_H = 2\pi f_H$ is upper cutoff frequency.
- Closed-Loop Frequency Transfer Function: $A_f(s) = [A_0 / (1 + s / \omega_H)] / [1 + \beta A_0 / (1 + s / \omega_H)] = A_0 / (1 + A_0 \beta + s / \omega_H)$.
- Dividing by $(1 + A_0 \beta)$: $A_f(s) = [A_0 / (1 + A_0 \beta)] / [1 + s / (\omega_H (1 + A_0 \beta))]$.
- Closed-Loop Cutoff Frequency: New upper 3-dB frequency $\omega_{Hf} = \omega_H (1 + A_0 \beta) \Rightarrow f_{Hf} = f_H (1 + A_0 \beta)$. Upper cutoff frequency is extended by $(1 + A_0 \beta)$!
- Closed-Loop Lower Cutoff Frequency: Lower 3-dB frequency is lowered: $f_{Lf} = f_L / (1 + A_0 \beta)$.
- Gain-Bandwidth Product Invariance: $A_{f0} f_{Hf} = [A_0 / (1 + A_0 \beta)] [f_H (1 + A_0 \beta)] = A_0 f_H = \text{Constant (GBP)}.$

Nonlinear Distortion & Noise Reduction Dynamics

- Harmonic Distortion Reduction: Transistor non-linear transfer functions generate output harmonic distortion D . Applying negative feedback reduces output harmonic distortion to $D_f = D / (1 + A * \beta)$.
- Linearity Enhancement: By suppressing second and third harmonic amplitudes, negative feedback transforms non-linear amplifiers into high-fidelity linear stages.
- Signal-to-Noise Ratio (SNR) Analysis: Consider noise voltage V_n injected inside an amplifier stage: $V_o = A_1 A_2 V_s + A_2 * V_n$.
- Closed-Loop Output with Noise: $V_{of} = (A_1 A_2 V_s) / (1 + A_1 A_2 \beta) + (A_2 V_n) / (1 + A_1 A_2 \beta)$.
- Noise Limitation: Negative feedback reduces output noise generated WITHIN the feedback loop, but CANNOT improve SNR for noise present at the primary input source (V_{ni})!

Mechanisms of Positive Feedback & Regenerative Action

- Positive Feedback Transfer Function: Feedback signal x_f is added in-phase at the mixer: $x_i = x_s + x_f = x_s + \beta * x_o$.
- Closed-Loop Gain Formulation: $A_f = A / (1 - A * \beta)$.
- Regenerative Gain Enhancement: For $0 < A * \beta < 1$, the denominator $(1 - A * \beta) < 1$, causing closed-loop gain A_f to EXCEED open-loop gain A (gain enhancement).
- Instability & Saturation Region: As loop gain $A * \beta$ approaches +1, closed-loop gain $A_f \rightarrow \infty$. The system becomes unstable; any microscopic noise pulse triggers exponential growth until circuit supply rails latch.
- Applications of Positive Feedback: Controlled positive feedback is utilized in oscillators (sinusoidal/waveform generation), Schmitt triggers (bistable multivibrators with hysteresis), and regenerative receivers.

Barkhausen Criterion for Sustained Oscillations

- Oscillator Definition: An electronic circuit that converts DC power into a periodic AC output signal without requiring any external AC input ($x_s = 0$).
- Barkhausen Condition 1 (Loop Gain Magnitude): The magnitude of the loop transmission around the feedback loop must equal unity at the desired frequency of oscillation f_0 : $|T(j\omega_0)| = |A(j\omega_0)\beta(j\omega_0)| = 1$.
- Barkhausen Condition 2 (Loop Phase Shift): The total phase shift around the feedback loop must equal zero degrees or an integer multiple of 360 degrees: $\angle(A\beta) = 2n\pi$ (where $n = 0, 1, 2, \dots$).
- Practical Oscillation Build-Up: To guarantee initial startup from thermal noise, practical designs set initial loop gain magnitude slightly higher: $|A\beta| \approx 1.05 - 1.1$.
- Amplitude Stabilization: Non-linear amplitude limiting (such as FET AGC circuits, thermistors, or diode clamps) reduces effective gain to $|A\beta| = 1$ precisely at steady state to prevent signal clipping.

Frequency Response & Stability Margins (Gain & Phase Margins)

- Potential Instability in Negative Feedback Amplifiers: At high frequencies, internal amplifier poles introduce phase lag. If phase lag reaches 180 degrees while $|A * \beta| = 1$, negative feedback turns into positive feedback!
- Gain Crossover Frequency (f_c): The frequency at which loop gain magnitude equals unity (0 dB): $|A(j\omega_c) \beta(j\omega_c)| = 1$ (0 dB).
- Phase Crossover Frequency (f_{180}): The frequency at which loop phase shift equals -180 degrees: $\angle(A(j\omega_{180}) \beta(j\omega_{180})) = -180$ deg.
- Phase Margin (PM): $PM = 180 \text{ deg} + \angle(A(j\omega_c) \beta(j\omega_c))$. For stable operation without ringing, $PM \geq 45$ deg (60 deg optimal).
- Gain Margin (GM): $GM = -20 \log_{10}(|A(j\omega_{180}) \beta(j\omega_{180})|)$. Indicates how much loop gain can increase before instability occurs (typically $GM \geq 6$ dB - 10 dB).

Quantitative Analysis of Negative Feedback Amplifier

- Problem Statement: An open-loop amplifier has midband voltage gain $A_0 = 10,000$ (80 dB), upper 3-dB frequency $f_H = 1$ kHz, and second-harmonic distortion $D = 10\%$. Calculate negative feedback properties with $\beta = 0.009$.
- Step 1: Calculate Loop Gain & Feedback Factor: Loop gain $T = A_0 \beta = 10,000 \cdot 0.009 = 90$. Amount of feedback $F = 1 + A_0 \beta = 1 + 90 = 91$ (39.18 dB).
- Step 2: Calculate Closed-Loop Gain (A_f): $A_f = A_0 / (1 + A_0 \beta) = 10,000 / 91 = 109.89$ (40.82 dB). Note: $1 / \beta = 1 / 0.009 = 111.11$.
- Step 3: Calculate Gain Sensitivity Reduction: If open-loop gain A drops by 20% ($dA/A = -0.20$), closed-loop gain variation is $dA_f / A_f = (-0.20) / 91 = -0.0022 = -0.22\%$!
- Step 4: Calculate Closed-Loop Bandwidth: $f_{Hf} = f_H (1 + A_0 \beta) = 1 \text{ kHz} \cdot 91 = 91 \text{ kHz}$.
- Step 5: Calculate Reduced Distortion: $D_f = D / (1 + A_0 \beta) = 10\% / 91 = 0.11\%$.

Quantitative Oscillator Design using Barkhausen Criterion

- Problem Statement: Design an RC Phase-Shift Oscillator core using a 3-stage RC ladder network in the feedback path. Determine the required open-loop amplifier gain A and frequency of oscillation f_0 .
- RC Ladder Feedback Transfer Function: $\beta(s) = (sRC)^3 / [(sRC)^3 + 6(sRC)^2 + 5(sRC) + 1]$.
- Step 1: Setting Imaginary Part to Zero for Phase Shift: Setting $s = j\omega$ and equating imaginary terms to zero gives: $-\omega^3 R^3 C^3 + 5\omega R^2 C^2 = 0$.
- Oscillation Frequency Derivation: $\omega_0^2 R^2 C^2 = 6 \Rightarrow \omega_0 = 1 / (\sqrt{6} RC) \Rightarrow f_0 = 1 / (2\pi \sqrt{6} RC)$.
- Step 2: Evaluating Feedback Factor at f_0 : Substituting ω_0 into $\beta(j\omega_0)$ yields $\beta(j\omega_0) = -1 / 29$.
- Step 3: Applying Barkhausen Gain Condition: For sustained oscillations, $-A\beta = 1 \Rightarrow A(-1/29) = -1 \Rightarrow A = +29$.
- Conclusion: The forward amplifier must provide an inverting gain of at least $A = -29$ (or non-inverting gain $+29$ with net 180 deg ladder phase shift) to sustain oscillations.

Comparative Performance Matrix: Negative vs. Positive Feedback

- Voltage Gain: Negative feedback reduces gain [$A_f = A / (1 + A \beta)$]; *Positive feedback increases gain* [$A_f = A / (1 - A \beta)$].
- Gain Stability: Negative feedback increases stability [$dA_f/A_f = (dA/A)/(1 + A * \beta)$]; Positive feedback decreases stability (leads to saturation/latching).
- System Bandwidth: Negative feedback expands bandwidth [$f_{Hf} = f_H (1 + A \beta)$]; Positive feedback narrows bandwidth.
- Nonlinear Distortion: Negative feedback suppresses distortion [$D_f = D / (1 + A * \beta)$]; Positive feedback increases distortion.
- Terminal Noise (Internal): Negative feedback reduces internal noise; Positive feedback amplifies noise.
- Primary Applications: Negative feedback is used in audio/RF linear amplifiers and operational amplifiers; Positive feedback is used in sinusoidal oscillators, relaxation oscillators, and Schmitt triggers.

Summary & Core Takeaways of Feedback Concepts

- Closed-Loop Transfer Function: $A_f = A / (1 \pm A * \beta)$. Negative feedback uses '+', Positive feedback uses '-'.
- Desensitization & Independence: Under heavy negative feedback ($A * \beta \gg 1$), closed-loop gain depends solely on passive components: $A_f \approx 1 / \beta$.
- Performance Enhancements: Negative feedback increases upper cutoff frequency ($f_{Hf} = f_H (1 \pm A * \beta)$), maintains constant Gain-Bandwidth Product, and lowers distortion ($D_f = D / (1 \pm A * \beta)$).
- Barkhausen Criteria: Sustained oscillation requires $|A * \beta| = 1$ and loop phase shift angle($A * \beta$) = 0 deg or 360 deg.
- Stability Margin Requirements: Amplifier stability requires positive Phase Margin (PM ≥ 45 deg) and Gain Margin (GM ≥ 6 dB) to prevent unwanted oscillations.